

# ECE 171A: Linear Control System Theory

## Lecture 23: Loop Shaping

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# Outline

Feedback design via loop shaping

Design examples

Summary

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## Loop shaping

**Loop shaping:** choose a compensator  $C(s)$  that gives a loop transfer function  $L(s) = P(s)C(s)$  with a desired shape. — **Trial and error procedure**

- ▶ **Example — Nyquist stability theorem:** To make an unstable system stable we simply have to bend the Nyquist curve away from the critical point  $s = -1 + i0$ .
- ▶ **Method 1** (backward): Determine a loop transfer function that gives a closed loop system with the desired properties and then compute the controller as  $C(s) = L(s)/P(s)$ . *Drawbacks:*
  - lead to controllers of high order
  - there are limits if the process transfer function  $P(s)$  has poles and zeros in the right half-plane,
- ▶ **Method 2:** (forward)
  - Start with the process transfer function  $P(s)$
  - Change its gain to obtain the desired bandwidth,
  - Add (stable) poles and zeros on  $C(s)$  until the desired shape is obtained.

## Design considerations

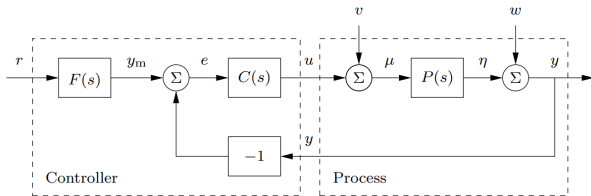


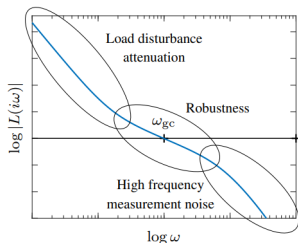
Figure: Block diagram of a control system with two degrees of freedom.

We need a suitable shape for the loop transfer function  $L(s) = P(s)C(s)$  that gives good **closed-loop performance** and good **stability margins**.

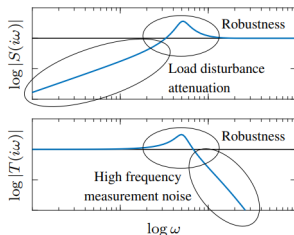
- ▶ Good performance requires that the loop transfer function  $L(s)$ 
  - is large for low frequencies — **good tracking of reference signals**
  - has **good attenuation** of low-frequency load disturbances.
- ▶ Since  $G_{yw} = S = 1/(1 + L(s))$  (note that  $G_{er} = S$  if  $F(s) = 1$ ), for frequencies  $\omega$  where  $|L(i\omega)| > 100$ 
  - disturbances will be attenuated by approximately a factor of 100
  - the steady-state tracking error  $|e(t)| = |r(t) - y(t)|$  is less than 1%.

## Design considerations

The loop transfer function should thus have roughly the shape shown in the following figure



(a) Gain plot of loop transfer function

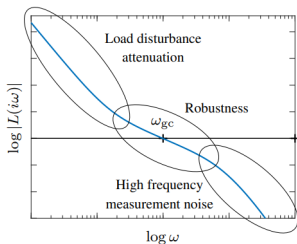


(b) Gain plot of sensitivity functions

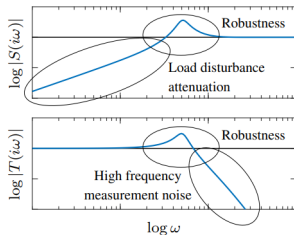
- ▶ It has unit gain at the gain crossover frequency ( $|L(i\omega_{gc})| = 1$ ),
- ▶ large gain for lower frequencies  $\omega < \omega_{gc}$ , and
- ▶ small gain for higher frequencies  $\omega > \omega_{gc}$

**Robustness** is determined by the shape of the loop transfer function around the gain-crossover frequency  $\omega_{gc}$ .

## Design considerations



(a) Gain plot of loop transfer function



(b) Gain plot of sensitivity functions

- ▶ It would be desirable to transition from high loop gain  $|L(i\omega)|$  at low frequencies to low loop gain as quickly as possible,
- ▶ **Robustness requirements** restrict how fast the gain can decrease:
  - For a minimum-phase system, the relationship between slope  $n_{gc}$  and phase margin  $\varphi_m$  (in degrees) is (no need to remember this equation)

$$n_{gc} \approx -2 + \frac{\varphi_m}{90}.$$

- ▶ Time delays and poles and zeros in the right half-plane impose further restrictions (Lecture 25/26)

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## Loop shaping via Lead and Lag Compensation

Loop shaping is a **trial-and-error** procedure.

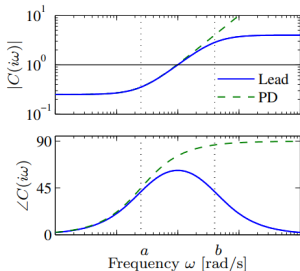
- ▶ Many specific procedures are available — they all require experience, but they also give **good insight** into the conflicting specifications.
- ▶ Start with a Bode plot of the process transfer function  $P(s)$
- ▶ Choose the gain crossover frequency  $\omega_{gc}$ 
  - A compromise between attenuation of load disturbances and injection of measurement noise.
- ▶ Attempt to shape the loop transfer function by changing the **controller gain** and **adding poles and zeros** to the controller transfer function.
  - the **loop gain at low frequencies** can be increased by so-called “**lag compensation**”
  - the **behavior around the crossover frequency** can be changed by so-called “**lead compensation**”.
- ▶ Different performance specifications are evaluated for each controller.

# Lead and Lag Compensation

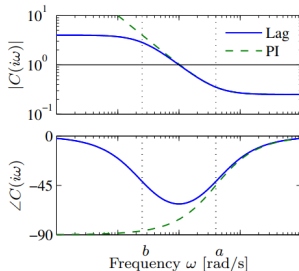
Simple compensators with transfer function

$$C(s) = k \frac{s + a}{s + b}, \quad a > 0, b > 0$$

- ▶ **Lag compensator** (Phase) if  $a > b$ ; a PI controller is a special case with  $b = 0$ .
- ▶ **Lead compensator** (Phase) if  $a < b$ ; a PD controller with filtering.

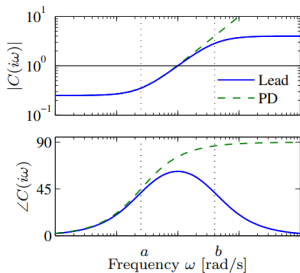


(a) Lead compensation,  $a < b$

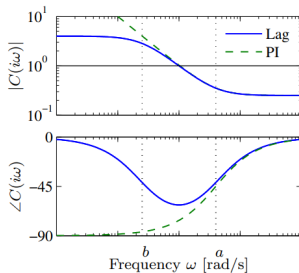


(b) Lag compensation,  $b < a$

# Lead and Lag Compensation



(a) Lead compensation,  $a < b$

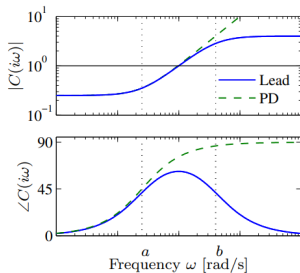


(b) Lag compensation,  $b < a$

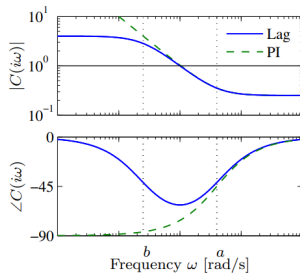
## General purpose of **Lag** compensation

- ▶ **increases the gain** at low frequencies
- ▶ **improve tracking performance** at low frequencies
- ▶ **improve disturbance attenuation** at low frequencies

# Lead and Lag Compensation



(a) Lead compensation,  $a < b$



(b) Lag compensation,  $b < a$

## General purpose of **Lead compensation**

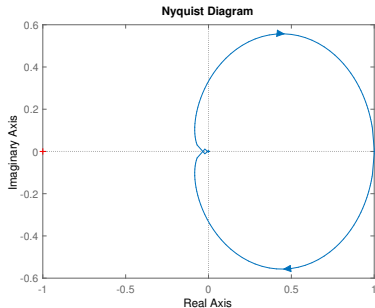
- ▶ Add **phase lead** in the frequency range between the pole and zero pair
- ▶ By appropriately choosing the location of this phase lead, we can provide **additional phase margin** at the gain crossover frequency.

## Example 1

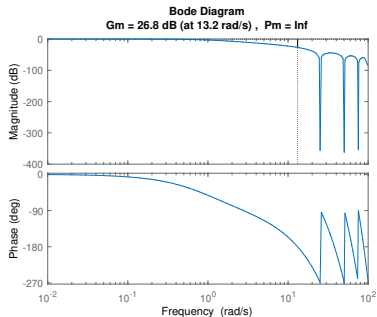
### Example (Example 12.4)

The transfer function for the system dynamics is

$$P(s) = \frac{a(1 - e^{-s\tau})}{s\tau(s + a)}, \quad a = 1, \quad \tau = 0.25$$



(a) Nyquist plot



(b) Bode plot with margins

## Example 1 - unit negative feedback

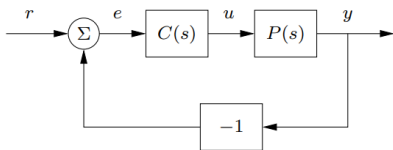
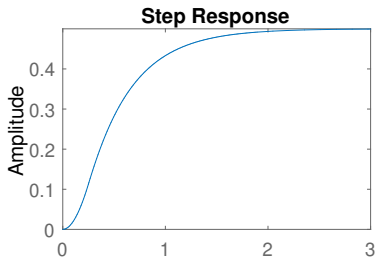
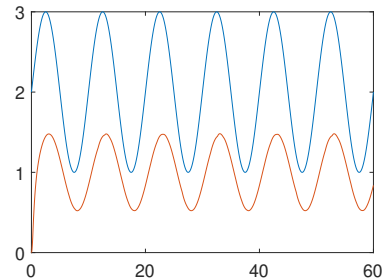


Figure: Unit negative feedback control  $C(s) = 1$

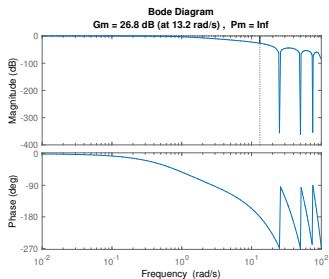


(a) Step response

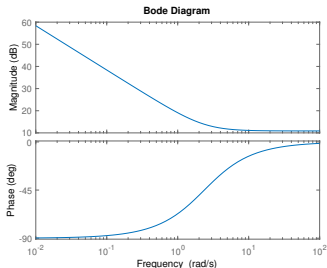


(b) Frequency response with  $T = 10$  seconds

# Example 1 - Lag compensation



(a)  $P(s)$



(b) Lag compensation (PI)

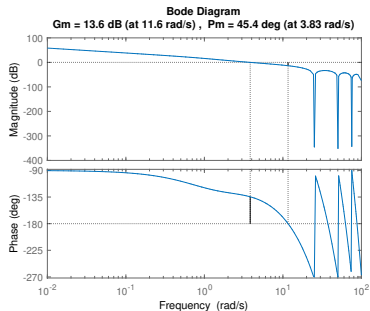


Figure: Margins for  $L(s) = P(s)C(s)$

$$C(s) = 3.5 + \frac{8.3}{s}$$

## Example 1 - Lag compensation

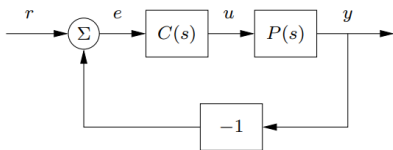
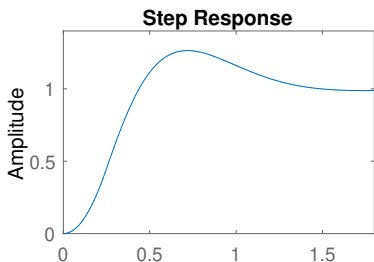
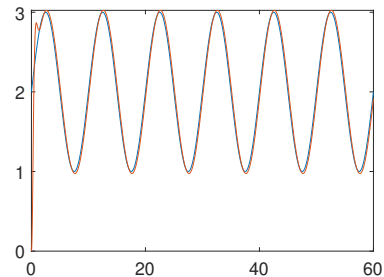


Figure: Feedback control with a lag compensator  $C(s) = k_p + \frac{k_i}{s}$



(a) Step response



(b) Frequency response with  $T = 10$  seconds



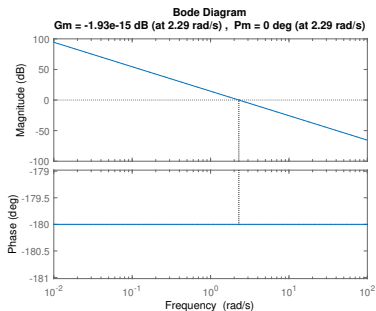
## Example 2

### Example (Example 12.5)

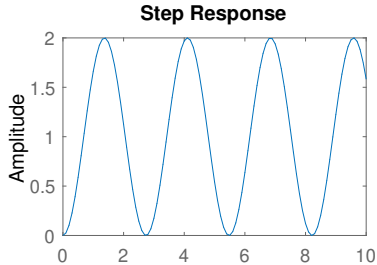
The transfer function for the system dynamics is

$$P(s) = \frac{r}{Js^2}, \quad r = 0.25, \quad J = 0.0475$$

- ▶ less than 1 % error in steady state;  $\leq 10\%$  tracking error up to 10 rad/s

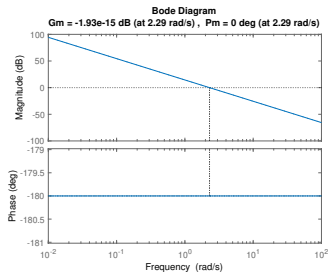


(a) Bode plot with margins

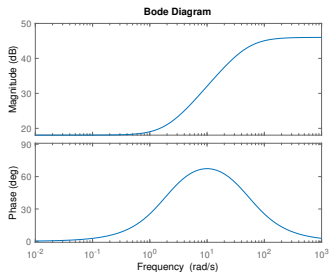


(b) Step response for unit negative feedback

## Example 2 - Lead compensation



(a)  $P(s)$



(b) Lead compensation

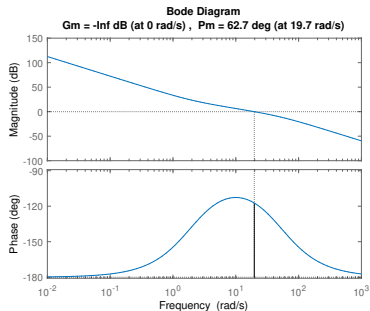
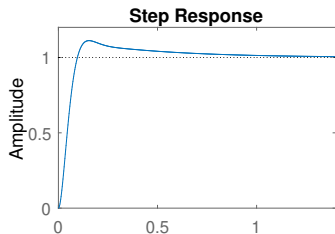


Figure: Margins for  $L(s) = P(s)C(s)$

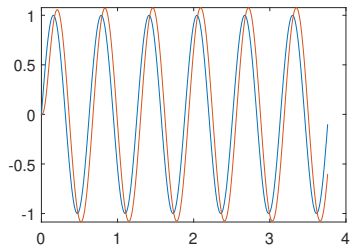
$$C(s) = k \frac{s + a}{s + b},$$

$$a = 2, b = 50, k = 200;$$

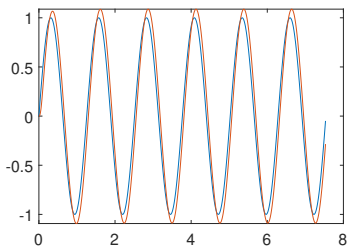
## Example 2 - time domain simulations



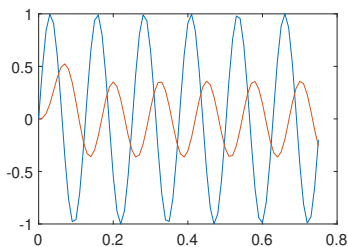
(a) Step response



(b) Frequency response  $\omega = 10$



(c) Frequency response  $\omega = 1$



(d) Frequency response  $\omega = 50$

# Outline

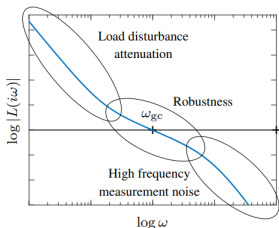
Feedback design via loop shaping

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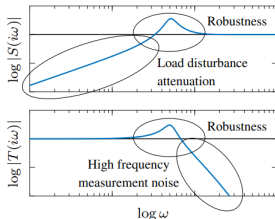
Summary

# Summary

- ▶ The **loop transfer function** should have roughly the shape below



(a) Gain plot of loop transfer function



(b) Gain plot of sensitivity functions

- ▶ General purpose of **Lag compensation**
  - increases the gain at low frequencies
  - improve tracking performance at low frequencies
  - improve disturbance attenuation at low frequencies
- ▶ General purpose of **Lead compensation**
  - Add phase lead in the frequency range between the pole and zero pair
  - By appropriately choosing the location of this phase lead, we can provide **additional phase margin** at the gain crossover frequency.