# Sample Complexity of Linear Quadratic Gaussian (LQG) Control for Output Feedback Systems 

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Model-based LQG Formulation
Model: a partially observed LTI system

$$
\begin{align*}
x_{t+1} & =A_{\star} x_{t}+B_{\star} u_{t}+B_{\star} w_{t}, \\
y_{t} & =C_{\star} x_{t}+v_{t} . \tag{1}
\end{align*}
$$

Optimal control problem:

$$
\begin{aligned}
\min _{u_{0}, u_{1}, \ldots \ldots} & \lim _{T \rightarrow \infty} \mathbb{E}\left[\frac{1}{T} \sum_{t=0}^{T}\left(y_{t}^{\top} Q y_{t}+u_{t}^{\top} R u_{t}\right)\right] \\
\text { subject to } & \text { (1). }
\end{aligned}
$$

Goal: Find optimal linear control policy

$$
\mathbf{u}(z)=\mathbf{K}(z) \mathbf{y}(z)
$$

minimizing (2).
Solution: Classical, e.g. observer/controller design with (model-based) Riccati equations.

> What if the model is completely unknown?
LQG - Unknown Model

Standard Assumption: Stable plant, i.e.,

$$
\rho\left(A_{\star}\right)<1 .
$$

## Proposed Design Procedure

I) Compute an estimate $\hat{\mathbf{G}}$ of the input/output impulse response

$$
\mathbf{G}_{\star}(z)=C_{\star}\left(z I-A_{\star}\right)^{-1} B_{\star}
$$

Let $\epsilon=\|\boldsymbol{\Delta}\|_{\infty}:=\left\|\mathbf{G}_{\star}-\hat{\mathbf{G}}\right\|_{\infty}$ be the estimation error.
III) Design a near-optimal controller..

- robustly stabilizing for all $\|\boldsymbol{\Delta}\|_{\infty} \leq \epsilon$


## Main contributions

> (1) End-to-end sample complexity bound on learning a robust LQG controller for open-loop stable plants.
> (2) Convex design of robust output-feedback controllers, enabled by the Input-Output Parametrization (IOP) [1].
> © LQG performance degrades linearly with the estimation error.

## Alternative Design Philosophies

- Idea 1: First estimate $\hat{A}, \hat{B}, \hat{C}$, then design optimal LQG controller (Certainty Equivalent Controller [2])

$$
\rightarrow \text { may not stabilize }\left(A^{\star}, B^{\star}, C^{\star}\right)!
$$

- Idea 2: So... First estimate

$$
\left\|\hat{A}-A_{\star}\right\|, \quad\left\|\hat{B}-B_{\star}\right\|, \quad\left\|\hat{C}-C_{\star}\right\|
$$

and then design a robustly stabilizing controller.
This is non-trivial!

- SLS from [3] can deal with state-feedback
- State-space estimation errors are only valid up to an unknown change of variables [2]
I) Nonasymptotic Identification
- Estimate Markov parameters

$$
\hat{\mathbf{G}}=\left[\begin{array}{lll}
\hat{C} \hat{B} & \hat{C} \hat{A} \hat{B} & \cdots \hat{C} \hat{A}^{T-1} \hat{B}
\end{array}\right]
$$

using standard least-squares identification

- Adapt nonasymptotic bound from [4]

$$
\begin{aligned}
\left\|\mathbf{G}_{\star}-\hat{\mathbf{G}}\right\|_{\infty} & \leq \frac{R_{w}+R_{v}+R_{e}}{\sigma_{u}} \sqrt{\frac{T}{N}} \\
& +\Phi\left(A_{\star}\right)\left\|C_{\star}\right\|\left\|B_{\star}\right\| \frac{\rho\left(A_{\star}\right)^{T}}{1-\rho\left(A_{\star}\right)}
\end{aligned}
$$

$\rightarrow \sqrt{\frac{T}{N}}$ : error 1) increases with modelling complexity $T$ and 2) decreases with more data $N$
II) Robust Controller Synthesis
(1) Input-Output Parametrization (IOP) [1]

> Optimize over K
$\equiv$
Optimize over closed-loop responses $(\mathbf{Y}, \mathbf{U}, \mathbf{W}, \mathbf{Z})$

$$
\left[\begin{array}{ll}
\mathbf{Y} & \mathbf{W} \\
\mathbf{U} \mathbf{Z}
\end{array}\right]:=\left[\begin{array}{cc}
\left(I-\mathbf{G}_{\star} \mathbf{K}\right)^{-1} & \left(I-\mathbf{G}_{\star} \mathbf{K}\right)^{-1} \mathbf{G}_{\star} \\
\mathbf{K}\left(I-\mathbf{G}_{\star} \mathbf{K}\right)^{-1} & \left(I-\mathbf{K G}_{\star}\right)^{-1}
\end{array}\right]
$$

(2) Tractable Robust Optimization ( $\triangle$ )

$$
\begin{aligned}
& \min _{\gamma \in[0,1 / \epsilon)} \frac{1}{1-\epsilon \gamma} \min _{\hat{\mathbf{Y}}, \hat{\mathbf{W}}, \hat{\mathbf{U}}, \hat{\mathbf{Z}}}\left\|\left[\begin{array}{cc}
\sqrt{1+h(\epsilon, \alpha)} \hat{\mathbf{Y}} & \hat{\mathbf{W}} \\
\hat{\mathbf{U}} & \hat{\mathbf{Z}}
\end{array}\right]\right\|_{\mathcal{H}_{2}} \\
& \text { subject to }\left[\begin{array}{ll}
I & -\hat{\mathbf{G}}
\end{array}\right]\left[\begin{array}{cc}
\hat{\mathbf{Y}} & \hat{\mathbf{W}} \\
\hat{\mathbf{U}} & \hat{\mathbf{Z}}
\end{array}\right]=\left[\begin{array}{ll}
I & 0
\end{array}\right] \text {, } \\
& {\left[\begin{array}{cc}
\hat{\mathbf{Y}} & \hat{\mathbf{W}} \\
\hat{\mathbf{U}} & \hat{\mathbf{Z}}
\end{array}\right]\left[\begin{array}{c}
-\hat{\mathbf{G}} \\
I
\end{array}\right]=\left[\begin{array}{l}
0 \\
I
\end{array}\right] \hat{\mathbf{Y}}, \hat{\mathbf{W}}, \hat{\mathbf{Z}} \in \mathcal{R} \mathcal{H}_{\infty},} \\
& \|\hat{\mathbf{U}}\|_{\infty} \leq \min (\gamma, \alpha) \text {. }
\end{aligned}
$$

## Remarks

- Program ( $\mathbf{\Delta}$ ) is quasi-convex
- Inner program is convex
- Golden-ratio search over $\gamma$
- Constraint on $\|\hat{\mathbf{U}}\|_{\infty}$ yields robustness [5]
- Controller $\hat{\mathbf{K}}=\hat{\mathbf{U}} \hat{\mathbf{Y}}^{-1}$ stabilizes all $\hat{\mathbf{G}}$ with
$\left\|\hat{\mathbf{G}}-\mathbf{G}_{\star}\right\|_{\infty} \leq \epsilon^{-}$
- $\alpha$ is an hyper-parameter to be tuned

Main Theorem - Sample Complexity

If $\epsilon$ is small enough, the controller $\mathbf{K}=\hat{\mathbf{U}}_{\star} \hat{\mathbf{Y}}_{\star}^{-1}$ which is optimal for ( $\mathbf{\Delta}$ ), is such that
i) $\mathbf{K}$ stabilizes the true plant $\mathbf{G}_{*}$,
ii) $\mathbf{K}$ introduces a suboptimality of at most

$$
\frac{J\left(\mathbf{G}_{\star}, \mathbf{K}\right)^{2}-J\left(\mathbf{G}_{\star}, \mathbf{K}_{\star}\right)^{2}}{J\left(\mathbf{G}_{\star}, \mathbf{K}_{\star}\right)^{2}}
$$

$\leq 20 \epsilon\left\|\mathbf{U}_{\star}\right\|_{\infty}+\mathcal{O}(\epsilon)$,
or, in terms of how much data is available: $\frac{J\left(\mathbf{G}_{\star}, \mathbf{K}\right)^{2}-J\left(\mathbf{G}_{\star}, \mathbf{K}_{\star}\right)^{2}}{J\left(\mathbf{G}_{\star}, \mathbf{K}_{\star}\right)^{2}} \sim \mathcal{O}\left(\frac{1}{\sqrt{N}}\right)$

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